

11MHZ CMOS Rail-to-Rail IO Opamps

Features

• Single-Supply Operation from +2.1V ~ +5.5V

• Rail-to-Rail Input / Output

• Gain-Bandwidth Product: 11MHz (Typ.)

• Low Input Bias Current: 1pA (Typ.)

Low Offset Voltage: 3.5mV (Max.)

• High Slew Rate: 9V/µs

Settling Time to 0.1% with 2V Step: 0.3μs

Low Noise: 8nV/ Hz @10kHz

Quiescent Current: 1.1mA per Amplifier (Typ.)

• Operating Temperature: -40°C ~ +125°C

• Small Package:

HGV8721 Available in SOT23-5, SOP-8 and SC70-5

Packages

HGV8722 Available in SOP-8 and MSOP-8 Packages
HGV8724 Available in SOP-14 and TSSOP-14 Packages

HGV8721N Available in SOT23-6 and SC70-6 Packages

General Description

The HGV872X have a high gain-bandwidth product of 11 MHz, a slew rate of 9V/ μs, and a quiescent current of 1.1mA per amplifier at 5V. The HGV8721N has a power-down disable feature that reduces the supply current to 90nA. The HGV872X are designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for HGV872X. They are specified over the extended indust rial temperature range (-40 °C to +125 °C). The operating range is from 2.1V to 5.5V. The HGV8721 single is available in Green SC70-5, SOT23-5 and SOP-8 packages. The HGV8722 dual is available in Green SOP-14 and TSSOP-14 packages. The HGV8721N single with shutdown is available in Green SOT23-6 and SC70-6 packages.

Applications

- Sensors
- Active Filters
- Cellular and Cordless Phones
- Laptops and PDAs

- Audio
- Handheld Test Equipment
- Battery-Powered Instrumentation
- A/D Converters

Pin Configuration

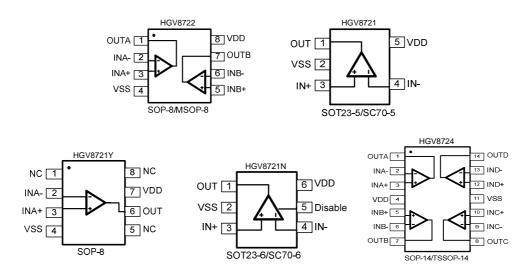


Figure 1. Pin Assignment Diagram



Absolute Maximum Ratings

Condition	Min	Max			
Power Supply Voltage (V _{DD} to Vss)	-0.5V	+7.5V			
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	V _{DD} +0.5V			
PDB Input Voltage	Vss-0.5V	+7V			
Operating Temperature Range	-40°C	+125°C			
Junction Temperature	+160)°C			
Storage Temperature Range	-55°C	+150°C			
Lead Temperature (soldering, 10sec)	+260°C				
Package Thermal Resistance (TA=+25°C)					
SOP-8, θ _{JA}	125°C/W				
MSOP-8, θ_{JA}	216°C/W				
SOT23-5, θ _{JA}	190°C/W				
SOT23-6, θ _{JA}	190°C/W				
SC70-5, θ _{JA}	333°C/W				
ESD Susceptibility					
НВМ	8KV				
MM	400V				

Note: Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.



Electrical Characteristics

(At Vs=5V, T_A = +25 °C, V_{CM} = $V_S/2$, R_L = 600 Ω , unless otherwise noted.)

PARAMETER	CONDITIONS	HGV8721/2/4/1N						
		TYP +25℃	MIN/MAX OVER TEMPERATURE					
			+25℃	0℃ to 70℃	-40 °C to 85 °C	-40 ℃ to	UNITS	MIN /
INPUT CHARACTERISTICS		1			II.	ı		<u>.1</u>
Input Offset Voltage (Vos)		0.8	3.5	3.9	4.3	4.6	mV	MAX
Input Bias Current (I _B)		1					pА	TYP
Input Offset Current (Ios)		1					pА	TYP
Input Common Mode Voltage Range (V _{CM})	V _S = 5.5V	-0.1 to +5.6					V	TYP
Common Mode Rejection Ratio (CMRR)	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 4V	82	65	64	64	63	dB	MIN
Common Mode regestion realis (OMITA)	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 5.6V	75					dB	MIN
Open-Loop Voltage Gain (A _{OL})	$R_L = 600\Omega, V_O = 0.15V \text{ to } 4.85V$	90	80	76	75	68	dB	MIN
1 1 0 (32)	$R_L = 10k\Omega, V_O = 0.05V \text{ to } 4.95V$	108					dB	MIN
Input Offset Voltage Drift ($\Delta V_{OS}/\Delta_T$)		2.4					μV/°C	TYP
OUTPUT CHARACTERISTICS					I			1
Output Voltage Swing from Rail	R _L = 600Ω	0.1					V	TYP
	$R_L = 10k\Omega$	0.015					V	TYP
Output Current (I _{OUT})		70	55	45	42	38	mA	MIN
Closed-Loop Output Impedance	f = 100kHz, G = 1	7.5					Ω	TYP
POWER-DOWN DISABLE								
Turn-On Time		1.1					μs	TYP
Turn-Off Time		0.3					μs	TYP
DISABLE Voltage-Off			0.8				V	MAX
DISABLE Voltage-On			2				V	MIN
POWER SUPPLY								
Operating Voltage Range			2.1	2.1	2.1	2.1	٧	MIN
Power Supply Rejection Ratio (PSRR)	V _S = +2.5V to +5.5V		5.5	5.5	5.5	5.5	V	MAX
	$V_{CM} = (-V_S) + 0.5V$	91	74	72	72	68	dB	MIN
Quiescent Current/Amplifier (I _Q)	I _{OUT} = 0	1.1	1.5	1.65	1.7	1.85	mA	MAX
Supply Current when Disabled		90					nA	MAX
(HGV8721N Only)								



Electrical Characteristics

(At Vs=5V, T_A = +25 °C, V_{CM} = $V_S/2$, R_L = 600 Ω , unless otherwise noted.)

PARAMETER	CONDITIONS	HGV8721/2/4/1N							
		TYP	MIN/MAX OVER TEMPERATURE						
		.05%	+25℃	0℃ to	-40℃ to	-40℃to	UNITS	MIN /	
		+25℃		70℃	85℃	125℃		MAX	
DYNAMIC PERFORMANCE			•						
Gain-Bandwidth Product (GBP)	R _L = 10kΩ, C _L = 100pF	11					MHz	TYP	
Phase Margin (φ _O)	$R_L = 10k\Omega, C_L = 100pF$	51					Degrees	TYP	
Full Power Bandwidth (BWP)	$<$ 1% distortion, R _L = 600 Ω	400					kHz	TYP	
Slew Rate (SR)	$G = +1$, 2V Step, $R_L = 10$ kΩ	9					V/µs	TYP	
Settling Time to 0.1% (t _S)	G = +1, 2V Step, R_L = 600Ω	0.3					μs	TYP	
Overload Recovery Time	V _{IN} ·Gain = VS, R _L = 600Ω	1.5					μs	TYP	
NOISE PERFORMANCE									
Voltage Noise Density (e _n)	f = 1kHz	11.5					nV/\sqrt{Hz}	TYP	
	f = 10kHz	8					nV/\sqrt{Hz}	TYP	

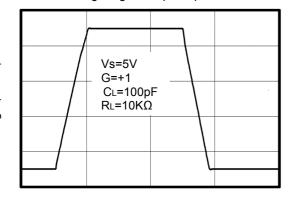


Voltage (1V/div)

Typical Performance characteristics

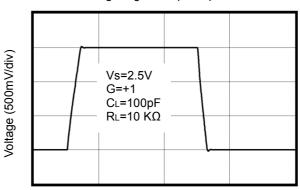
(At Vs=5V, TA = +25 $^{\circ}\mathrm{C}$, VcM = Vs/2, RL = 600 $\!\Omega$, unless otherwise noted.)

Large-Signal Step Response



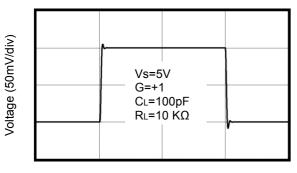
Time (1µs/div)

Large-Signal Step Response



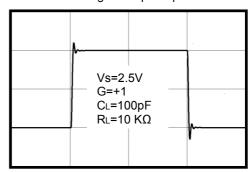
Time (1µs/div)

Small-Signal Step Response



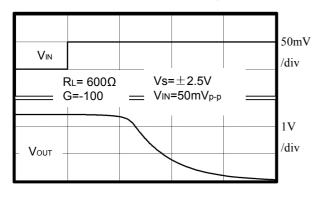
Time (1µs/div)

Small-Signal Step Response



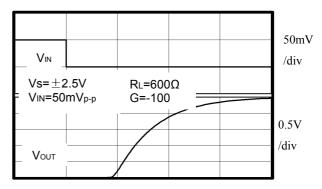
Time (1µs/div)

Positive Overload Recovery



Time (2µs/div)

Negative Overload Recovery



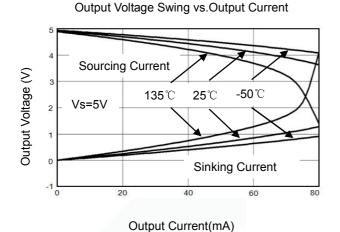
Time (2µs/div)

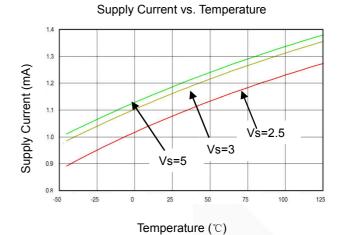
Voltage (50mV/div)

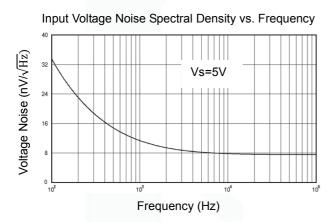


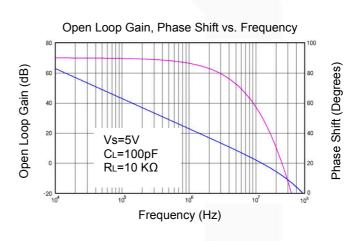
Typical Performance characteristics

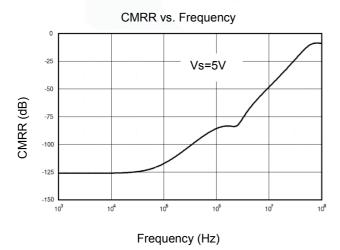
(At Vs=5V, TA = +25 $^{\circ}$ C, VcM = Vs/2, RL = 600 Ω , unless otherwise noted.)

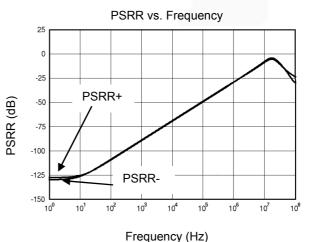














Application Note

Size

HGV872X series op amps are unity-gain stable and sui table for a wide range of general-purpose applications. The small footprints of the HGV872X series packages save space on printed circuit boards and enable the design of smaller electronic products.

Power Supply Bypassing and Board Layout

HGV872X series operates from a single 2.1V to 5.5V supply or dual ± 1.05 V to ± 2.75 V supplies. For best performance, a 0.1μ F ceramic capacitor should be placed close to the V_{DD} pin in single supply operation. For dual supply operation, both V_{DD} and V_{SS} supplies should be bypassed to ground with separate 0.1μ F ceramic capacitors.

Low Supply Current

The low supply current (typical 1.1mA per channel) of HGV872X series will help to maximize battery life. They are ideal for battery powered systems

Operating Voltage

HGV872X series operate under wide input supply voltage (2.1V to 5.5V). In addition, all temperature specifications apply from -40 °C to +125 °C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime

Rail-to-Rail Input

The input common-mode range of HGV872X series extends 100mV beyond the supply rails (V $_{SS}$ -0.1V to V $_{DD}$ +0.1V). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of HGV872X series can typically swing to less than 2mV from supply rail in light resistive loads (>100k Ω), and 15mV of supply rail in moderate resistive loads (10k Ω).

Capacitive Load Tolerance

The HGV872X family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

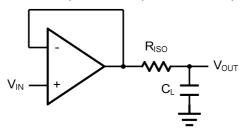


Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the R_{ISO} resistor value, the more stable V_{OUT} will be. However, if there is a resistive load R_L in parallel with the capacitive load, a voltage divider (proportional to R_{ISO}/R_L) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2. R_F provides the DC accuracy by feed-forward the V_{IN} to R_L. C_F



and $R_{\rm ISO}$ serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of C_F . This in turn will slow down the pulse response.

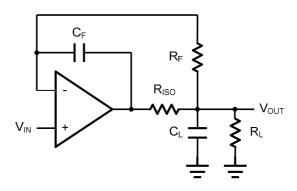


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy



Typical Application Circuits

Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using HGV872X.

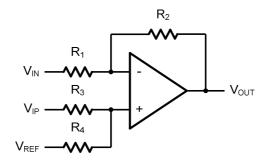


Figure 4. Differential Amplifier

$$V_{\text{OUT}} = (\frac{R_1 + R_2}{R_2 + R_4}) \frac{R_4}{R_1} V_{\text{IN}} - \frac{R_2}{R_1} V_{\text{IP}} + (\frac{R_1 + R_2}{R_2 + R_4}) \frac{R_3}{R_1} V_{\text{REF}}$$

If the resistor ratios are equal (i.e. R₁=R₃ and R₂=R₄), then

$$V_{\text{OUT}} = \frac{R_2}{R_1} (V_{\text{IP}} - V_{\text{IN}}) + V_{\text{REF}}$$

Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by $-R_2/R_1$. The filter has a -20dB/decade roll-off after its corner frequency $f_c=1/(2\pi R_3C_1)$.

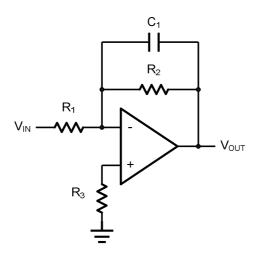


Figure 5. Low Pass Active Filter



Instrumentation Amplifier

The triple HGV872X can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of R_2/R_1 . The two differential voltage followers assure the high input impedance of the amplifier.

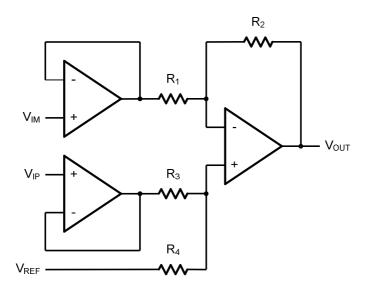


Figure 6. Instrument Amplifier